



Research article

Event-triggered stabilization for networked control systems under random occurring deception attacks

Dong Xu¹, Xinling Li¹, Weipeng Tai^{1, 2,*} and Jianping Zhou²

¹ Research Institute of Information Technology, Anhui University of Technology, Ma'anshan 243000, China

² School of Computer Science & Technology, Anhui University of Technology, Ma'anshan 243032, China

* **Correspondence:** Email: taiweipeng@ahut.edu.cn.

Abstract: This paper copes with event-triggered stabilization for networked control systems subject to deception attacks. A new switched event-triggered scheme (ETS) is designed by introducing a term regarding the last triggering moment in the trigger condition. This increases the difficulty of triggering, thus reducing trigger times compared to some existing ETSs. Furthermore, to cater for actual deception attack behavior, the occurrence of deception attacks is assumed to be a time-dependent stochastic variable that obeys the Bernoulli distribution with probability uncertainty. By means of a piecewise-defined Lyapunov function, a sufficient condition is developed to assure that the close-loop system under deception attacks is exponentially stable in regards to mean square. On the basis of this, a joint design of the desired trigger and feedback-gain matrices is presented. Finally, a simulation example is given to confirm the validity of the design method.

Keywords: networked control systems; deception attacks; lyapunov function; stabilization; event-triggered control

1. Introduction

Networked control systems (NCSs) have been widely studied and applied in many fields in the past few decades, including DC motor [1], intelligent transportation [2], teleoperation [3], etc. Compared with traditional feedback control systems, the components in NCSs transmit data packets through communication networks, which reduce wiring requirements, save installation costs, and improve the maintainability [4]. However, public and open communication networks are vulnerable to malicious attacks by hackers. Cyber attacks can not only disrupt data transmission but also indirectly cause controlled plants to malfunction and stop operating by injecting fake data [5]. For example, in 2011, the

“Stuxnet” virus invaded Iran’s Bolshevik nuclear power plant, causing massive damage to its nuclear program [6]; in 2015, the “BlackEnergy” trojan virus successfully attacked Ukraine’s power companies and cut off the local power supply [7]. Therefore, the security of NCS under cyber attacks has received increasing attention, and many illuminating results have been reported (see surveys [8, 9]).

Current cyber-attacks may be roughly divided into two categories: denial of service (DoS) attacks and deception attacks. DoS attacks can cause network congestion or paralysis by sending a large amount of useless request information, thereby blocking the transmission of signals [10, 11]. Compared to the simple and direct blocking of signal transmission in DoS attacks, deception attacks are more subtle and difficult to detect. More specifically, deception attacks compromise data integrity by hijacking sensors to tamper with measurement data and control signals. In some cases, they may be more damaging than DoS attacks [12]. Therefore, many researchers have conducted analysis and synthesis of NCSs under deception attacks. For instance, Du et al. [13] presented stability conditions for wireless NCSs subjected to deception attacks on the data link layer, and determined the maximum permissible deception attack time. Hu et al. [14] examined discrete-time stochastic NCSs with deception attacks and packet loss, and gave security analysis and controller design. In [15], Gao et al. investigated nonlinear NCSs faced with several types of deception attacks, and proposed asynchronous observer design strategies. Notably, the existing literature characterized the random occurrence of deception attacks by introducing a stochastic variable that obeys the Bernoulli distribution with a fixed probability, which is over-limited in reality.

In previous literature on NCSs, control tasks are often executed in a fixed/variable period, giving rise to the so-called sampled-data control (time-triggered control) [16–19]. Although this control scheme is simple and easy to implement, it has two shortcomings. On the one hand, to stabilize the system in some extreme scenarios, it is necessary to set a small sampling interval, which results in a large number of redundant data packets and may cause network congestion. On the other hand, due to the lack of a judgment condition for the system state, even if the stabilization of the system is achieved, the sampling task will not stop, which is undoubtedly a waste of network resources. In contrast to the time-triggered scheme (TTS), the event-triggered scheme (ETS) can significantly conquer these shortcomings and save communication resources [20–22]. In the ETS, the execution frequency of the control task is limited by introducing an event generator. That is, only when the change in system state exceeds a given threshold, an event will be generated and the controller will execute [23–25]. Based on the above facts, a multitude of studies have been conducted on event-triggered control for NCSs suffering from deception attacks (see [26–30]). However, most ETS designs under deception attacks adopted the periodic event-triggered scheme (PETS). The PETS can avoid the Zeno phenomenon by periodically checking the event trigger conditions but inevitably lose some system state information. To resolve this contradiction, Selivanov et al. proposed a switched event-triggered scheme (SETS) [31]. By switching between TTS and continuous event triggering [32], the SETS not only avoids the Zeno phenomenon, but also fully utilizes the system information, thereby further reducing the number of event triggers [33–36].

Inspired by the above facts, the subject of this study is SETS-based stabilization for NCSs under random deception attacks. By means of the Lyapunov function method, a sufficient condition is developed to assure that the close-loop system is mean square exponentially (MSE) stable. Then, a co-design of the trigger and feedback-gain matrices for the SETS-based controller is presented in terms of linear matrix inequalities (LMIs), which can be checked easily. The major contributions can be

outlined as follows: 1) A new SETS (NSETS), which additionally introduces a term concerning the last event triggering time, is designed. In contrast to the SETS in [31], the NSETS can further reduce the number of trigger times while maintaining performance. 2) Compared with the existing literature (see, e.g., [28–30]), the occurrence of deception attacks under our consideration is assumed to be a time-dependent stochastic variable that obeys the Bernoulli distribution with probability uncertainty, which is more realistic. 3) Stability analysis criterion and computationally tractable controller design strategy are presented by utilizing a piecewise-defined Lyapunov function together with a few matrix inequalities.

Notation. In this paper, we use $\mathbb{R}^{m \times n}$ and \mathbb{R}^m to denote $m \times n$ -dimensional real matrices and m -dimensional Euclidean space, respectively. For each $Z \in \mathbb{R}^{m \times n}$, we use $Z < 0$ ($Z \leq 0$) to indicate that the matrix Z is real symmetric negative definite (semi-negative definite), $\mathcal{S}\{Z\}$ to denote the sum of $Z + Z^T$, and $\lambda_M(Z)$ and $\lambda_m(Z)$ to represent the maximum eigenvalue and the minimum eigenvalue of the symmetric matrix Z , respectively. We use $\|\cdot\|$, \mathcal{P}_r , and \mathcal{E} to represent the Euclidean norm, probability operator, and expectation operator, respectively. In addition, we use $\text{diag}\{\cdot\}$ to denote a diagonal matrix and $*$ to denote a symmetric term in a symmetric matrix.

2. Preliminaries

In this section, we first give the system description, deception-attack model, and NSETS design, and then formulate the switched closed-loop system.

Consider a linear time-invariant system (LTIS) described as

$$\dot{x}(t) = Ax(t) + Bu(t). \quad (2.1)$$

In system (2.1), A and B are known system matrices of suitable dimensions, $x(t) \in \mathbb{R}^{n_x}$, $u(t) \in \mathbb{R}^{n_u}$, denote the state, control input, respectively.

In the event-triggered NCS, an event generator is used to determine whether the lately sampled data should be forwarded to the controller. When cyber-attacks do not happen, any state signals that satisfy the triggering condition can be received successfully by the controller, and the corresponding control signals can be received successfully by the actuator. However, when the communication network is subject to deception attacks, the control signal will be substituted with fake data, the performance of the controller will inevitably decline, and even lead to system instability in some extreme cases.

In view of the openness of network protocols, it is quite possible that attackers can inject the information transmitted in NCSs with fake data. In this paper, we suppose that the deception signal $d(x(t))$ satisfying Assumption 1 will totally substitute the original transmission data.

Assumption 1. [26] The signal of deception attack $d(x(t))$ satisfies

$$\|d(x(t))\| \leq \|Dx(t)\|, \quad (2.2)$$

where D is a real constant matrix.

To demonstrate the randomness of deception attacks, a Bernoulli variable $\beta(t) \in \{0, 1\}$ with uncertain probability is introduced:

$$\mathcal{P}_r(\beta(t) = 1) = \alpha + \Delta\alpha(t), \quad \mathcal{P}_r(\beta(t) = 0) = 1 - \alpha - \Delta\alpha(t). \quad (2.3)$$

Clearly, $\mathcal{E}\{\beta(t)\} = \alpha + \Delta\alpha(t)$. Then, the signal transmitted under deception attacks can be presented as

$$\tilde{x}(t) = (1 - \beta(t))d(x(t)) + \beta(t)x(t).$$

Remark 2.1. Motivated by [26–30], the deception attacks are assumed to be randomly occurring. However, different from these references, this paper supposes that the probability of deception attack occurrence allows for some uncertainty; that is, there is a small positive constant $\bar{\alpha}$ such that $\|\Delta\alpha(t)\| \leq \bar{\alpha}$.

Taking into consideration the constrained network resources, a NSETS is devised as follows:

$$\begin{aligned} t_{\sigma+1} &= \min\{t \geq t_{\sigma} + h \mid [x(t) - x(t_{\sigma})]^T M[x(t) - x(t_{\sigma})] \\ &> \varepsilon_1 x^T(t) M x(t) + \varepsilon_2 x^T(t_{\sigma}) M x(t_{\sigma})\}, \end{aligned} \quad (2.4)$$

where t_{σ} is the last triggering moment with $\sigma \in \mathbb{N}$, $M > 0$ is a trigger matrix to be determined, $h > 0$ and ε_i ($i \in \{1, 2\}$) are predefined parameters that stand for the waiting interval and triggering thresholds, respectively.

Remark 2.2. Inspired by [31, 33–36], the SETS is used in this paper. However, unlike the existing SETSs, the proposed NSETS introduces an additional term $\varepsilon_2 x^T(t_{\sigma}) M x(t_{\sigma})$ into the trigger condition to more efficiently utilize the state information at the current moment and previous triggering moment. The introduced $\varepsilon_2 x^T(t_{\sigma}) M x(t_{\sigma})$ increases the difficulty of the event triggering and, therefore, prolongs the interval between triggered events.

In accordance with the proposed NSETS, the control input of system (2.1) under deception attacks with uncertain probability can be designed as

$$u(t) = K(1 - \beta(t))d(x(t)) + K\beta(t)x(t_{\sigma}), \quad (2.5)$$

where K is the feedback gain that remains to be determined.

Then, we introduce $\mathcal{T}_{\sigma}^1 = [t_{\sigma}, t_{\sigma} + h)$ and $\mathcal{T}_{\sigma}^2 = [t_{\sigma} + h, t_{\sigma+1})$, under event-triggered controller (2.5), system (2.1) can be characterized as

$$\begin{cases} \dot{x}(t) = [A + BK\beta(t)]x(t) + BK(1 - \beta(t))d(x(t)) \\ \quad - BK\beta(t) \int_{t_{\sigma}}^t \dot{x}(s)ds, t \in \mathcal{T}_{\sigma}^1, \\ \dot{x}(t) = [A + BK\beta(t)]x(t) + BK(1 - \beta(t))d(x(t)) \\ \quad + BK\beta(t)e(t), t \in \mathcal{T}_{\sigma}^2, \end{cases} \quad (2.6)$$

where $\sigma \in \mathbb{N}$ and $e(t) = x(t_{\sigma}) - x(t)$ satisfying

$$e^T(t) M e(t) \leq \varepsilon_1 x^T(t) M x(t) + \varepsilon_2 x^T(t_{\sigma}) M x(t_{\sigma}). \quad (2.7)$$

Remark 2.3. Unlike general ETSs, the controlled system with the SETS needs to switch between periodic sampling and continuous event triggering to ensure the performance of the system. Inspired by [31], we denote the sampling error and triggering error by $\int_{t_{\sigma}}^t \dot{x}(s)ds$ and $e(t)$, respectively, and rewrite the controlled system as switched system (2.6) with two different modes.

Next, we introduce a definition of mean square exponential stability and several lemmas.

Definition 2.1 ([37]). System (2.6) is said to be MSE stable if there exist two scalars $c_1 > 0$ and $c_2 > 0$ such that

$$\mathcal{E}\{\|x(t)\|\} \leq c_1 e^{-c_2 t} \mathcal{E}\{\|x_0\|\}, \quad \forall t \geq 0,$$

Lemma 2.2 ([38]). For any positive definite matrix $S \in \mathbb{R}^{n \times n}$, scalars a, b ($a > b$), and a vector function $\eta : [a, b] \rightarrow \mathbb{R}^n$, we have

$$\left[\int_b^a \eta(\delta) d\delta \right]^T S \left[\int_b^a \eta(\delta) d\delta \right] \leq (a - b) \int_b^a \eta^T(\delta) S \eta(\delta) d\delta.$$

Lemma 2.3 ([39]). For given $a, b \in \mathbb{R}^n$, and positive definite matrix $Q \in \mathbb{R}^{n \times n}$, we have

$$2a^T b \leq a^T Q a + b^T Q^{-1} b.$$

Lemma 2.4 ([40]). For a prescribed constant $\vartheta > 0$ and real matrices Π, X_i, Y_i , and Z_i ($i = 1, \dots, n$), if

$$\begin{bmatrix} \Pi & X_Y \\ * & Z \end{bmatrix} < 0$$

holds, where $X_Y = [X_1 + \vartheta Y_1, X_2 + \vartheta Y_2, \dots, X_n + \vartheta Y_n]$ and $Z = \text{diag}\{\mathcal{S}\{-\vartheta Z_1\}, \mathcal{S}\{-\vartheta Z_2\}, \dots, \mathcal{S}\{-\vartheta Z_n\}\}$, then we have

$$\Pi + \sum_{i=1}^n \mathcal{S}\{X_i Z_i^{-1} Y_i^T\} < 0. \quad (2.8)$$

Now, the issue of event-triggered control in response to deception attacks can be expressed more specifically as follows: given a LTIS in (2.1), determine the NSETS-based controller in (2.5) such that, for all admissible deception attacks $d(x(t))$, the switched closed-loop system in (2.6) is MSE stable.

3. Main results

In this section, we first establish the exponential stability criterion for system (2.1), then develop a joint design method for the trigger matrix and feedback gain.

3.1. Stability analysis

To be consistent with the switched closed-loop system (2.6), we construct the following Lyapunov function:

$$V(t) = \begin{cases} V_1(t) = V_p(t) + V_q(t) + V_u(t), & t \in \mathcal{T}_\sigma^1, \\ V_2(t) = V_p(t), & t \in \mathcal{T}_\sigma^2 \end{cases} \quad (3.1)$$

where

$$V_p(t) = x^T(t) P x(t),$$

$$V_q(t) = (t_\sigma + h - t) \int_{t_\sigma}^t e^{-2\theta(t-s)} \dot{x}^T(s) Q \dot{x}(s) ds,$$

$$V_u(t) = (t_\sigma + h - t) \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix}^T \begin{bmatrix} \frac{\mathcal{S}\{U_1\}}{2} & -U_1 + U_2 \\ * & \mathcal{S}\{U_2 - \frac{U_1}{2}\} \end{bmatrix} \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix},$$

and P, Q are symmetric positive matrices, U_1, U_2 are general matrices, and θ is a positive constant.

Based on Lyapunov function (3.1), an exponential stability criterion of system (2.6) can be established, which is provided as follows:

Theorem 3.1. For given positive scalars $\varepsilon_1, \varepsilon_2, h, \theta, \alpha, \bar{\alpha}$ and matrices K, D , under the NSETS (2.4) and random deception attacks, system (2.6) is MSE stable, if there exist symmetric matrices $P > 0, Q > 0, M > 0, \Lambda_i > 0 (i \in \{1, 2, 3, 4\})$, and general matrices $U_1, U_2, W_1, W_2, W_3, N_{1,1}, N_{2,1}, N_{1,2}, N_{2,2}$, such that

$$\Sigma = \begin{bmatrix} P + h\mathcal{S}\left\{\frac{U_1}{2}\right\} & -hU_1 + hU_2 \\ * & -h\mathcal{S}\left\{U_2 - \frac{U_1}{2}\right\} \end{bmatrix} > 0, \quad (3.2)$$

$$\Sigma_1 = \begin{bmatrix} \Omega_{11} & \Omega_{12} \\ * & \Omega_3 \end{bmatrix} < 0, \quad (3.3)$$

$$\Sigma_2 = \begin{bmatrix} \Omega_{21} & \Omega_{22} \\ * & \Omega_3 \end{bmatrix} < 0, \quad (3.4)$$

$$\Sigma_3 = \begin{bmatrix} \Omega_{31} & \Omega_{32} \\ * & \Omega_{33} \end{bmatrix} < 0 \quad (3.5)$$

hold true, where

$$\Omega_{11} = \begin{bmatrix} \Omega_{1,1}^1 & \Omega_{1,2}^1 & \Omega_{1,3}^1 & \Omega_{1,4}^1 \\ * & \Omega_{2,2}^1 & \Omega_{2,3}^1 & \Omega_{2,4}^1 \\ * & * & \Omega_{3,3}^1 & 0 \\ * & * & * & -I \end{bmatrix} + \bar{\alpha}^2 \mathcal{R}_1^T (\Lambda_1 + \Lambda_2) \mathcal{R}_1,$$

$$\mathcal{R}_1 = [I \ 0 \ 0 \ -I],$$

$$\Omega_{21} = \begin{bmatrix} \Omega_{1,1}^2 & \Omega_{1,2}^2 & \Omega_{1,3}^2 & \Omega_{1,4}^2 & \Omega_{1,5}^2 \\ * & -\mathcal{S}\{N_{2,1}\} & W_2^T & \Omega_{2,4}^2 & \Omega_{2,5}^2 \\ * & * & \Omega_{3,3}^2 & 0 & hW_3^T \\ * & * & * & -I & 0 \\ * & * & * & * & \Omega_{5,5}^2 \end{bmatrix} + \bar{\alpha}^2 \mathcal{R}_2^T (\Lambda_1 + \Lambda_2) \mathcal{R}_2,$$

$$\mathcal{R}_2 = [I \ 0 \ 0 \ -I \ -hI], \mathcal{R}_3 = [I \ 0 \ -I \ I],$$

$$\Omega_{31} = \begin{bmatrix} \Omega_{1,1}^3 & \Omega_{1,2}^3 & \Omega_{1,3}^3 & \Omega_{1,4}^3 \\ * & -\mathcal{S}\{N_{2,2}\} & \Omega_{2,3}^3 & \Omega_{2,4}^3 \\ * & * & \Omega_{3,3}^3 & 0 \\ * & * & * & -I \end{bmatrix} + \bar{\alpha}^2 \mathcal{R}_3^T (\Lambda_3 + \Lambda_4) \mathcal{R}_3,$$

$$\Omega_{12} = \begin{bmatrix} \Omega_{1,6}^T & 0 & 0 & 0 \\ 0 & \Omega_{2,7}^T & 0 & 0 \end{bmatrix}^T, \Omega_{32} = \begin{bmatrix} \Omega_{1,5}^{3T} & 0 & 0 & 0 \\ 0 & \Omega_{2,6}^{3T} & 0 & 0 \end{bmatrix}^T,$$

$$\begin{aligned}
\Omega_{22} &= \begin{bmatrix} \Omega_{12} \\ 0 \end{bmatrix}, \quad \Omega_3 = \begin{bmatrix} -\Lambda_1 & 0 \\ 0 & -\Lambda_2 \end{bmatrix}, \quad \Omega_{33} = \begin{bmatrix} -\Lambda_3 & 0 \\ 0 & -\Lambda_4 \end{bmatrix}, \\
\Omega_{1,1}^1 &= \mathcal{S} \left\{ N_{1,1}^T (A + BK\alpha) - W_1 + (2\theta h - 1) \frac{U_1}{2} \right\} + 2\theta P + D^T D, \\
\Omega_{1,1}^2 &= \Omega_{1,1}^1 - \theta h \mathcal{S}\{U_1\}, \\
\Omega_{1,2}^1 &= P - W_2 - N_{1,1}^T + \frac{h}{2} \mathcal{S}\{U_1\} + (A + BK\alpha)^T N_{2,1}, \\
\Omega_{1,2}^2 &= \Omega_{1,2}^1 - \frac{h}{2} \mathcal{S}\{U_1\}, \\
\Omega_{1,3}^1 &= W_1^T - W_3 + (1 - 2\theta h)(U_1 - U_2), \quad \Omega_{1,3}^2 = \Omega_{1,3}^1 + 2\theta h(U_1 - U_2), \\
\Omega_{1,4} &= (1 - \alpha)N_{1,1}^T BK, \quad \Omega_{1,5} = h(W_1^T - \alpha N_{1,1}^T BK), \quad \Omega_{1,6} = N_{1,1}^T BK, \\
\Omega_{2,2}^1 &= hQ - \mathcal{S}\{N_{2,1}\}, \quad \Omega_{2,3}^1 = W_2^T - h(U_1 - U_2), \\
\Omega_{2,4} &= (1 - \alpha)N_{2,1}^T BK, \quad \Omega_{2,5} = h(W_2^T - \alpha N_{2,1}^T BK), \\
\Omega_{2,7} &= N_{2,1}^T BK, \quad \Omega_{5,5} = -he^{-2\theta h} Q, \\
\Omega_{3,3}^1 &= \mathcal{S} \left\{ W_3 + (2\theta h - 1) \left(\frac{U_1}{2} - U_2 \right) \right\}, \\
\Omega_{3,3}^2 &= \Omega_{3,3}^1 - \mathcal{S} \left\{ 2\theta h \left(\frac{U_1}{2} - U_2 \right) \right\}, \\
\Omega_{1,1}^3 &= 2\theta P + \mathcal{S} \left\{ N_{1,2}^T (A + BK\alpha) \right\} + (\varepsilon_1 + \varepsilon_2)M + D^T D, \\
\Omega_{1,2}^3 &= P - N_{1,2}^T + (A + BK\alpha)^T N_{2,2}, \quad \Omega_{1,3}^3 = N_{1,2}^T BK\alpha + \varepsilon_2 M, \\
\Omega_{1,4}^3 &= N_{1,2}^T BK(1 - \alpha), \quad \Omega_{1,5}^3 = N_{1,2}^T BK, \quad \Omega_{2,3}^3 = \alpha N_{2,2}^T BK, \\
\Omega_{2,4}^3 &= (1 - \alpha)N_{2,2}^T BK, \quad \Omega_{2,6}^3 = N_{2,2}^T BK, \quad \Omega_{3,3}^3 = (\varepsilon_2 - 1)M.
\end{aligned}$$

Proof. Since

$$\begin{aligned}
&V_p(t) + V_u(t) \\
&= x^T(t)Px(t) + (t_\sigma + h - t) \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix}^T \begin{bmatrix} \frac{\mathcal{S}\{U_1\}}{2} & -U_1 + U_2 \\ * & \mathcal{S}\{\frac{U_1}{2} - U_2\} \end{bmatrix} \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix} \\
&= \frac{t_\sigma + h - t}{h} \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix}^T \Sigma \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix} + \frac{t - t_\sigma}{h} \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix}^T \begin{bmatrix} P & 0 \\ * & 0 \end{bmatrix} \begin{bmatrix} x(t) \\ x(t_\sigma) \end{bmatrix}, \tag{3.6}
\end{aligned}$$

which, in conjunction with $P > 0$ and $\Sigma > 0$, ensures the positive definiteness of $V_p(t) + V_u(t)$. From (3.1) and (3.6), we have

$$\min \{ \lambda_m(P), \lambda_m(\Sigma) \} \|x(t)\|^2 \leq V(t). \tag{3.7}$$

For any $t \in \mathcal{T}_\sigma^1$, $\sigma \in \mathbb{N}$, differentiating $V_1(t)$ along the trajectories of the system (2.6) and taking mathematical expectations on it yields:

$$\mathcal{E}\{\dot{V}_1(t)\} = \mathcal{E}\{\dot{V}_p(t)\} + \mathcal{E}\{\dot{V}_q(t)\} + \mathcal{E}\{\dot{V}_u(t)\},$$

where

$$\mathcal{E}\{\dot{V}_p(t)\} = \mathcal{E}\{-2\theta V_p(t)\} + 2\theta x^T(t)Px(t) + 2x^T(t)P\dot{x}(t),$$

$$\begin{aligned}
\mathcal{E}\{\dot{V}_q(t)\} &= \mathcal{E}\{-2\theta V_q(t)\} - \int_{t_\sigma}^t e^{-2\theta(t-s)} \dot{x}^T(s) Q \dot{x}(s) ds \\
&\quad + (t_\sigma + h - t) \dot{x}^T(t) Q \dot{x}(t), \\
\mathcal{E}\{\dot{V}_u(t)\} &= \mathcal{E}\{-2\theta V_u(t)\} \\
&\quad + (t_\sigma + h - t) [\dot{x}^T(t) \mathcal{S}\{U_1\} x(t) + 2\dot{x}^T(t) (-U_1 + U_2) x(t_\sigma)] \\
&\quad + [2\theta(t_\sigma + h - t) - 1] \times \left[x^T(t) \frac{\mathcal{S}\{U_1\}}{2} x(t) \right. \\
&\quad \left. + 2\dot{x}^T(t) (-U_1 + U_2) x(t_\sigma) + x^T(t_\sigma) \mathcal{S}\left\{\frac{U_1}{2} - U_2\right\} x(t_\sigma) \right].
\end{aligned}$$

Thus

$$\begin{aligned}
\mathcal{E}\{\dot{V}_1(t)\} &\leq \mathcal{E}\{-2\theta V_1(t)\} + 2\theta x^T(t) P x(t) + 2x^T(t) P \dot{x}(t) + (t_\sigma + h - t) \dot{x}^T(t) Q \dot{x}(t) \\
&\quad - e^{-2\theta h} \int_{t_\sigma}^t \dot{x}^T(s) Q \dot{x}(s) ds + (t_\sigma + h - t) [\dot{x}^T(t) \mathcal{S}\{U_1\} x(t) \\
&\quad + 2\dot{x}^T(t) (-U_1 + U_2) x(t_\sigma)] + [2\theta(t_\sigma + h - t) - 1] \\
&\quad \times \left[x^T(t) \frac{\mathcal{S}\{U_1\}}{2} x(t) + 2\dot{x}^T(t) (-U_1 + U_2) x(t_\sigma) \right. \\
&\quad \left. + x^T(t_\sigma) \mathcal{S}\left\{\frac{U_1}{2} - U_2\right\} x(t_\sigma) \right]. \tag{3.8}
\end{aligned}$$

Denote

$$\phi(t) = \frac{1}{t - t_\sigma} \int_{t_\sigma}^t \dot{x}(s) ds,$$

where the case that $\phi(t)|_{t=t_\sigma}$ can be understood as $\lim_{t \rightarrow t_\sigma} \phi(t) = \dot{x}(t_\sigma)$ [41]. Then, utilizing Lemma 2.2 gives

$$- \int_{t_\sigma}^t \dot{x}(s) Q \dot{x}(s) ds \leq -(t - t_\sigma) \phi^T(t) Q \phi(t). \tag{3.9}$$

Furthermore, using the Newton-Leibniz formula and the expectation of (2.6), we can write

$$\begin{aligned}
0 &= 2 \left[x^T(t) W_1^T + \dot{x}^T(t) W_2^T + x^T(t_\sigma) W_3^T \right] \\
&\quad \times [-x(t) + x(t_\sigma) + (t - t_\sigma) \phi(t)], \tag{3.10}
\end{aligned}$$

$$\begin{aligned}
0 &= 2 \left[x^T(t) N_{1,1}^T + \dot{x}^T(t) N_{2,1}^T \right] [-\dot{x}(t) + A x(t) + BK(\alpha + \Delta\alpha(t)) x(t) \\
&\quad + BK(1 - \alpha - \Delta\alpha(t)) d(x(t)) - (t - t_\sigma) BK(\alpha + \Delta\alpha(t)) \phi(t)]. \tag{3.11}
\end{aligned}$$

Setting $\zeta_1(t) = \text{col}\{x(t), \dot{x}(t), x(t_\sigma), d(x(t))\}$, $\zeta_2(t) = \text{col}\{x(t), \dot{x}(t), x(t_\sigma), d(x(t)), \phi(t)\}$, by employing Lemma 2.3, for positive definite matrices Λ_1 and Λ_2 , we can write the following inequalities:

$$\begin{aligned}
&2x^T(t) N_{1,1}^T BK \Delta\alpha(t) [x(t) - d(x(t)) - (t - t_\sigma) \phi(t)] \\
&= 2x^T(t) N_{1,1}^T BK \Delta\alpha(t) \left[\frac{t_\sigma + h - t}{h} \mathcal{R}_1 \zeta_1(t) + \frac{t - t_\sigma}{h} \mathcal{R}_2 \zeta_2(t) \right]
\end{aligned}$$

$$\begin{aligned} &\leq \frac{t_\sigma + h - t}{h} \bar{\alpha}^2 \mathcal{S}_1^T(t) \mathcal{R}_1^T \Lambda_1 \mathcal{R}_1 \mathcal{S}_1(t) + \frac{t - t_\sigma}{h} \bar{\alpha}^2 \mathcal{S}_2^T(t) \mathcal{R}_2^T \Lambda_1 \mathcal{R}_2 \mathcal{S}_2(t) \\ &\quad + x^T(t) N_{1,1}^T B K \Lambda_1^{-1} K^T B^T N_{1,1} x(t). \end{aligned} \quad (3.12)$$

Similarly,

$$\begin{aligned} &2\dot{x}^T(t) N_{2,1}^T B K \Delta \alpha(t) [x(t) - d(x(t)) - (t - t_\sigma) \phi(t)] \\ &\leq \frac{t_\sigma + h - t}{h} \bar{\alpha}^2 \mathcal{S}_1^T(t) \mathcal{R}_1^T \Lambda_2 \mathcal{R}_1 \mathcal{S}_1(t) + \frac{t - t_\sigma}{h} \bar{\alpha}^2 \mathcal{S}_2^T(t) \mathcal{R}_2^T \Lambda_2 \mathcal{R}_2 \mathcal{S}_2(t) \\ &\quad + \dot{x}^T(t) N_{2,1}^T B K \Lambda_2^{-1} K^T B^T N_{2,1} \dot{x}(t). \end{aligned} \quad (3.13)$$

Moreover, by using Assumption 1, we can derive

$$x^T(t) D^T D x(t) - d^T(x(t)) d(x(t)) \geq 0. \quad (3.14)$$

By using inequalities (3.8)–(3.14), we have

$$\begin{aligned} &\mathcal{E}\{\dot{V}_1(t)\} + 2\theta \mathcal{E}\{V_1(t)\} \\ &\leq \frac{t_\sigma + h - t}{h} \mathcal{S}_1^T(t) \Omega_{11} \mathcal{S}_1(t) + \frac{t - t_\sigma}{h} \mathcal{S}_2^T(t) \Omega_{21} \mathcal{S}_2(t) \\ &\quad + x^T(t) N_{1,1}^T B K \Lambda_1^{-1} K^T B^T N_{1,1} x(t) + \dot{x}^T(t) N_{2,1}^T B K \Lambda_2^{-1} K^T B^T N_{2,1} \dot{x}(t), \end{aligned}$$

which, in conjunction with the Schur complement, $\Sigma_1 < 0$, and $\Sigma_2 < 0$, guarantees that

$$\mathcal{E}\{\dot{V}_1(t)\} + 2\theta \mathcal{E}\{V_1(t)\} \leq 0. \quad (3.15)$$

Denote $\xi_3(t) = \text{col}\{x(t), \dot{x}(t), e(t), d(x(t))\}$. Then, in a similar way to the above proof, it is not hard to obtain

$$\begin{aligned} \mathcal{E}\{\dot{V}_2(t)\} &\leq -2\theta \mathcal{E}\{V_2(t)\} + 2\theta x^T(t) P x(t) + x^T(t) D^T D x(t) \\ &\quad + 2x^T(t) P \dot{x}(t) + 2[x^T(t) N_{1,2}^T + \dot{x}^T(t) N_{2,2}^T] \\ &\quad \times [-\dot{x}(t) + A x(t) + B K(\alpha + \Delta \alpha(t)) x(t) \\ &\quad + B K(1 - \alpha - \Delta \alpha(t)) d(x(t)) \\ &\quad + B K(\alpha + \Delta \alpha(t)) e(t)] + \varepsilon_1 x^T(t) M x(t) \\ &\quad + \varepsilon_2 [e(t) + x(t)]^T M [e(t) + x(t)] \\ &\quad - e^T(t) M e(t) - d^T(x(t)) d(x(t)). \end{aligned} \quad (3.16)$$

By using Lemma 2.3, for given positive definite matrices Λ_3 and Λ_4 , we can obtain

$$\begin{aligned} &2x^T(t) N_{1,2}^T B K \Delta \alpha(t) \mathcal{S}_3(t) \\ &\leq x^T(t) N_{1,2}^T B K \Lambda_3^{-1} K^T B^T N_{1,2} x(t) + \bar{\alpha}^2 \mathcal{S}_3^T(t) \mathcal{R}_3^T \Lambda_3 \mathcal{R}_3 \mathcal{S}_3(t), \end{aligned} \quad (3.17)$$

$$\begin{aligned} &2\dot{x}^T(t) N_{2,2}^T B K \Delta \alpha(t) \mathcal{S}_3(t) \\ &\leq \dot{x}^T(t) N_{2,2}^T B K \Lambda_4^{-1} K^T B^T N_{2,2} \dot{x}(t) + \bar{\alpha}^2 \mathcal{S}_3^T(t) \mathcal{R}_3^T \Lambda_4 \mathcal{R}_3 \mathcal{S}_3(t). \end{aligned} \quad (3.18)$$

Combining (3.16)–(3.18), we get

$$\mathcal{E}\{\dot{V}_2(t)\} + 2\theta \mathcal{E}\{V_2(t)\} \leq \mathcal{S}_3^T(t) \Sigma_3 \mathcal{S}_3(t) + x^T(t) N_{1,2}^T B K \Lambda_3^{-1} K^T B^T N_{1,2} x(t)$$

$$+\dot{x}^T(t)N_{2,2}^T BK\Lambda_4^{-1}K^T B^T N_{2,2}\dot{x}(t)$$

for any $t \in \mathcal{T}_\sigma^2$, which together with the Schur complement and $\Omega_{31} < 0$, implies that

$$\mathcal{E}\{\dot{V}_2(t)\} + 2\theta\mathcal{E}\{V_2(t)\} \leq 0. \quad (3.19)$$

According to the expression of $V(t)$, it is easy to obtain that

$$\begin{aligned} V_q(t_\sigma) &= V_u(t_\sigma) = 0, \\ \lim_{t \rightarrow (t_\sigma+h)^-} V_q(t) &= \lim_{t \rightarrow (t_\sigma+h)^-} V_u(t) = 0, \end{aligned}$$

which confirms the continuity of $V(t)$ at instants t_σ and $t_\sigma + h$.

Then combining (3.15) and (3.19), for $t \in \mathcal{T}_\sigma^1$ we can derive

$$\begin{aligned} \mathcal{E}\{V(t)\} &\leq \mathcal{E}\{V(t_\sigma)\}e^{-2\theta(t-t_\sigma)} \\ &\leq \mathcal{E}\{V(t_{\sigma-1})\}e^{-2\theta(t-t_{\sigma-1})} \\ &\dots \\ &\leq \mathcal{E}\{V(0)\}e^{-2\theta t}. \end{aligned}$$

Similarly, for $t \in \mathcal{T}_\sigma^2$, the same results can be obtained. In the light of (3.1), we get

$$\mathcal{E}\{V(0)\} \leq \lambda_M(P)\mathcal{E}\{\|x(0)\|^2\}.$$

It can be concluded that for any $t \in \mathcal{T}_\sigma^1 \cup \mathcal{T}_\sigma^2$,

$$\mathcal{E}\{V(t)\} \leq e^{-2\theta t}\mathcal{E}\{V(0)\},$$

which, together with (3.7), gives

$$\mathcal{E}\{\|x(t)\|\} \leq \sqrt{\frac{\min\{\lambda_m(P), \lambda_m(\Sigma)\}}{\lambda_M(P)}} e^{-\theta t}\mathcal{E}\{\|x_0\|\}.$$

This completes the proof.

When there is no deception attacks, $d(x(t)) = 0$, and system (2.6) becomes

$$\begin{cases} \dot{x}(t) = (A + BK)x(t) - BK \int_{t_\sigma}^t \dot{x}(s)ds, t \in \mathcal{T}_\sigma^1, \\ \dot{x}(t) = (A + BK)x(t) + BK e(t), t \in \mathcal{T}_\sigma^2, \end{cases} \quad (3.20)$$

and we can write the following sufficient condition:

Corollary 1. For given feedback gain matrix K and positive scalars $\varepsilon_1, \varepsilon_2, h, \theta$, under the NSETS (2.4), system (3.20) is exponentially stable, if there exist symmetric matrices $P > 0, Q > 0, M > 0$, and general matrices $U_1, U_2, W_1, W_2, W_3, N_{1,1}, N_{2,1}, N_{1,2}, N_{2,2}$, such that (3.2) and the following LMIs

$$\begin{bmatrix} \Omega_{1,1}^1 & \Omega_{1,2}^1 & \Omega_{1,3}^1 \\ * & \Omega_{2,2}^1 & \Omega_{2,3}^1 \\ * & * & \Omega_{3,3}^1 \end{bmatrix} < 0, \quad (3.21)$$

$$\begin{bmatrix} \Omega_{1,1}^2 & \Omega_{1,2}^2 & \Omega_{1,3}^2 & \Omega_{1,5} \\ * & -\mathcal{S}\{N_{2,1}\} & W_2^T & \Omega_{2,5} \\ * & * & \Omega_{3,3}^2 & hW_3^T \\ * & * & * & \Omega_{5,5} \end{bmatrix} < 0, \quad (3.22)$$

$$\begin{bmatrix} \Omega_{1,1}^3 & \Omega_{1,2}^3 & \Omega_{1,3}^3 \\ * & -\mathcal{S}\{N_{2,2}\} & \Omega_{2,3}^3 \\ * & * & \Omega_{3,3}^3 \end{bmatrix} < 0 \quad (3.23)$$

hold true, where matrix blocks such as $\Omega_{1,1}^1$, $\Omega_{1,2}^1$, and $\Omega_{1,3}^1$ are the same as those in Theorem 3.1 except that $\alpha = 1$ and $\bar{\alpha} = 0$.

3.2. Event-triggered controller synthesis

In this section, we will explore the feasibility of the event-triggered controller design. On the basis of Theorem 3.1, the trigger matrix M and feedback-gain matrix K can be derived from the following result:

Theorem 3.2. For given positive scalars ε_1 , ε_2 , h , θ , α , $\bar{\alpha}$, ϑ and matrix D , under the NSETS (2.4) and random deception attacks, switched system (2.6) with control gain $K = X^{-1}Y$ is MSE stable, if there exist symmetric matrices $P > 0$, $Q > 0$, $U > 0$, $M > 0$, $\bar{\Lambda}_i > 0$ ($i \in \{1, 2, 3, 4\}$), and general matrices X , Y , \tilde{W}_1 , \tilde{W}_2 , \tilde{W}_3 , $\tilde{N}_{1,1}$, $\tilde{N}_{2,1}$, $\tilde{N}_{1,2}$, $\tilde{N}_{2,2}$, such that (3.2) and the following LMIs

$$\begin{bmatrix} \tilde{\Omega}_{11} & \tilde{\Omega}_{12} \\ * & \tilde{\Omega}_3 \end{bmatrix} < 0, \quad (3.24)$$

$$\begin{bmatrix} \tilde{\Omega}_{21} & \tilde{\Omega}_{22} \\ * & \tilde{\Omega}_3 \end{bmatrix} < 0, \quad (3.25)$$

$$\begin{bmatrix} \tilde{\Omega}_{31} & \tilde{\Omega}_{32} \\ * & \tilde{\Omega}_3 \end{bmatrix} < 0 \quad (3.26)$$

hold true, where

$$\begin{aligned} \tilde{\Omega}_{11} &= \begin{bmatrix} \tilde{\Omega}_1^1 & \tilde{\Omega}_2^1 \\ * & \tilde{\Omega}_3^1 \end{bmatrix}, \quad \tilde{\Omega}_{21} = \begin{bmatrix} \tilde{\Omega}_1^2 & \tilde{\Omega}_2^2 \\ * & \tilde{\Omega}_3^1 \end{bmatrix}, \quad \tilde{\Omega}_{31} = \begin{bmatrix} \tilde{\Omega}_1^3 & \tilde{\Omega}_2^3 \\ * & \tilde{\Omega}_3^3 \end{bmatrix}, \\ \tilde{\Omega}_1^1 &= \begin{bmatrix} \tilde{\Omega}_{1,1}^1 & \tilde{\Omega}_{1,2}^1 & \tilde{\Omega}_{1,3}^1 & \tilde{\Omega}_{1,4}^1 \\ * & \tilde{\Omega}_{2,2}^1 & \tilde{\Omega}_{2,3}^1 & \tilde{\Omega}_{2,4}^1 \\ * & * & \tilde{\Omega}_{3,3}^1 & 0 \\ * & * & * & -I \end{bmatrix} + \bar{\alpha}^2 \tilde{\mathcal{R}}_1^T (\tilde{\Lambda}_1 + \tilde{\Lambda}_2) \tilde{\mathcal{R}}_1, \\ \tilde{\mathcal{R}}_1 &= [I \ 0 \ 0 \ -I], \\ \tilde{\Omega}_1^2 &= \begin{bmatrix} \tilde{\Omega}_{1,1}^2 & \tilde{\Omega}_{1,2}^2 & \tilde{\Omega}_{1,3}^2 & \tilde{\Omega}_{1,4} & \tilde{\Omega}_{1,5} \\ * & \tilde{\Omega}_{2,2}^2 & \tilde{W}_2^T & \tilde{\Omega}_{2,4} & \tilde{\Omega}_{2,5} \\ * & * & \tilde{\Omega}_{3,3}^2 & 0 & h\tilde{W}_3^T \\ * & * & * & -I & 0 \\ * & * & * & * & -he^{-2\theta h} Q \end{bmatrix} + \bar{\alpha}^2 \tilde{\mathcal{R}}_2^T (\tilde{\Lambda}_1 + \tilde{\Lambda}_2) \tilde{\mathcal{R}}_2, \\ \tilde{\mathcal{R}}_2 &= [I \ 0 \ 0 \ -I \ -hI], \quad \tilde{\mathcal{R}}_3 = [I \ 0 \ -I \ I], \end{aligned}$$

$$\begin{aligned}
\tilde{\Omega}_1^3 &= \begin{bmatrix} \tilde{\Omega}_{1,1}^3 & \tilde{\Omega}_{1,2}^3 & \tilde{\Omega}_{1,3}^3 & \tilde{\Omega}_{1,4}^3 \\ * & -\mathcal{S}\{\tilde{N}_{2,2}\} & \alpha G^T Y & \tilde{\Omega}_{2,4}^3 \\ * & * & (\varepsilon_2 - 1)M & 0 \\ * & * & * & -I \end{bmatrix} + \bar{\alpha}^2 \tilde{\mathcal{R}}_3^T (\tilde{\Lambda}_3 + \tilde{\Lambda}_4) \tilde{\mathcal{R}}_3, \\
\tilde{\Omega}_2^1 &= \begin{bmatrix} Y^T G & 0 & 0 & 0 \\ 0 & Y^T G & 0 & 0 \end{bmatrix}^T, \quad \tilde{\Omega}_2^2 = \begin{bmatrix} \tilde{\Omega}_2^1 \\ 0 \end{bmatrix}, \\
\tilde{\Omega}_2^3 &= \begin{bmatrix} Y^T G & 0 & 0 & 0 \\ 0 & Y^T G & 0 & 0 \end{bmatrix}^T, \quad \tilde{\Omega}_3^1 = \begin{bmatrix} -\tilde{\Lambda}_1 & 0 \\ 0 & -\tilde{\Lambda}_2 \end{bmatrix}, \\
\tilde{\Omega}_3^3 &= \begin{bmatrix} -\tilde{\Lambda}_3 & 0 \\ 0 & -\tilde{\Lambda}_4 \end{bmatrix}, \quad \tilde{\Omega}_3 = -\vartheta \begin{bmatrix} X & 0 \\ 0 & X \end{bmatrix} - \vartheta \begin{bmatrix} X^T & 0 \\ 0 & X^T \end{bmatrix}, \\
\tilde{\Omega}_{12} &= \begin{bmatrix} \tilde{\Omega}_{1,8}^T & 0 & 0 & (1-\alpha)\vartheta Y & \vartheta Y & 0 \\ \alpha\vartheta Y & \tilde{\Omega}_{2,9}^T & 0 & (1-\alpha)\vartheta Y & 0 & \vartheta Y \end{bmatrix}^T, \\
\tilde{\Omega}_{22} &= \begin{bmatrix} \tilde{\Omega}_{1,8}^T & 0 & 0 & (1-\alpha)\vartheta Y & -\alpha\vartheta h Y & \vartheta Y & 0 \\ \alpha\vartheta Y & \tilde{\Omega}_{2,9}^T & 0 & (1-\alpha)\vartheta Y & -\alpha\vartheta h Y & 0 & \vartheta Y \end{bmatrix}^T, \\
\tilde{\Omega}_{32} &= \begin{bmatrix} \tilde{\Omega}_{1,8}^{3T} & 0 & \alpha\vartheta Y & (1-\alpha)\vartheta Y & \vartheta Y & 0 \\ \alpha\vartheta Y & \tilde{\Omega}_{2,9}^{3T} & \alpha\vartheta Y & (1-\alpha)\vartheta Y & 0 & \vartheta Y \end{bmatrix}^T, \\
\tilde{\Omega}_{1,1}^1 &= 2\theta P + D^T D + \mathcal{S} \left\{ \tilde{N}_{1,1}^T A - \tilde{W}_1 + (2\theta h - 1) \frac{U_1}{2} + \alpha G^T Y \right\}, \\
\tilde{\Omega}_{1,1}^2 &= \tilde{\Omega}_{1,1}^1 - \theta h \mathcal{S}\{U_1\}, \\
\tilde{\Omega}_{1,2}^1 &= P - \tilde{W}_2 - \tilde{N}_{1,1}^T + \frac{h}{2} \mathcal{S}\{U_1\} + A^T \tilde{N}_{2,1}, \\
\tilde{\Omega}_{1,2}^2 &= \tilde{\Omega}_{1,2}^1 - \frac{h}{2} \mathcal{S}\{U_1\}, \\
\tilde{\Omega}_{1,3}^1 &= \tilde{W}_1^T - \tilde{W}_3 + (1 - 2\theta h)(U_1 - U_2), \\
\tilde{\Omega}_{1,3}^2 &= \tilde{\Omega}_{1,3}^1 + 2\theta h(U_1 - U_2), \\
\tilde{\Omega}_{1,4} &= (1 - \alpha)G^T Y, \quad \tilde{\Omega}_{1,5} = h(\tilde{W}_1^T - \alpha G^T Y), \\
\tilde{\Omega}_{1,8} &= \tilde{N}_{1,1}^T B - G^T X + \alpha\vartheta Y^T, \quad \tilde{\Omega}_{2,2}^1 = hQ - \mathcal{S}\{\tilde{N}_{2,1}\}, \\
\tilde{\Omega}_{2,2}^2 &= -\mathcal{S}\{\tilde{N}_{2,1}\}, \quad \tilde{\Omega}_{2,3}^1 = \tilde{W}_2^T - h(U_1 - U_2), \\
\tilde{\Omega}_{2,4} &= (1 - \alpha)G^T Y, \\
\tilde{\Omega}_{2,5} &= h(\tilde{W}_2^T - \alpha G^T Y), \quad \tilde{\Omega}_{2,9} = \tilde{N}_{2,1}^T B - G^T X, \\
\tilde{\Omega}_{3,3}^1 &= \mathcal{S} \left\{ \tilde{W}_3 + (2\theta h - 1) \left(\frac{U_1}{2} - U_2 \right) \right\}, \\
\tilde{\Omega}_{3,3}^2 &= \tilde{\Omega}_{3,3}^1 - \mathcal{S} \left\{ 2\theta h \left(\frac{U_1}{2} - U_2 \right) \right\}, \\
\tilde{\Omega}_{1,1}^3 &= 2\theta P + \mathcal{S} \left\{ \tilde{N}_{1,2}^T A + \alpha G^T Y \right\} + (\varepsilon_1 + \varepsilon_2)M + D^T D, \\
\tilde{\Omega}_{1,2}^3 &= P - \tilde{N}_{1,2}^T + A^T \tilde{N}_{2,2} + \alpha Y^T G, \\
\tilde{\Omega}_{1,3}^3 &= \varepsilon_2 M + \alpha G^T Y, \\
\tilde{\Omega}_{1,4}^3 &= (1 - \alpha)G^T Y, \quad \tilde{\Omega}_{1,8}^3 = \tilde{N}_{1,2}^T B - G^T X + \alpha\vartheta Y^T,
\end{aligned}$$

$$\tilde{\Omega}_{2,4}^3 = (1 - \alpha)G^T Y, \quad \tilde{\Omega}_{2,9}^3 = \tilde{N}_{2,2}^T B - G^T X,$$

and G is an all-ones matrix with the same dimension as B .

Proof. By Lemma 2.4, it follows from (3.24) that

$$\tilde{\Omega}_{11} + \mathcal{S} \left\{ \begin{array}{cc} \left[\begin{array}{cc} \tilde{N}_{1,1}^T B - G^T L & 0 \\ 0 & \tilde{N}_{2,1}^T B - G^T L \end{array} \right] & \\ \left[\begin{array}{cc} X & 0 \\ 0 & X \end{array} \right]^{-1} \left[\begin{array}{cccccc} \alpha Y & 0 & 0 & (1 - \alpha)Y & Y & 0 \\ \alpha Y & 0 & 0 & (1 - \alpha)Y & 0 & Y \end{array} \right] & \end{array} \right\} < 0,$$

which can be equivalently rewritten as

$$\begin{bmatrix} \Omega_{11} & \Omega_{12} \\ * & \Omega_3 \end{bmatrix} < 0.$$

In other words, we can derive (3.3) from (3.24). Similarly, (3.4)–(3.5) can be obtained from (3.25)–(3.26). The proof is complete.

Corollary 2. For given positive scalars $\varepsilon_1, \varepsilon_2, h, \theta, \vartheta$, under the NSETS (2.4), switched system (3.20) with control gain $K = X^{-1}Y$ is exponentially stable, if there exist symmetric matrices $P > 0, Q > 0, U > 0, M > 0$, and general matrices $X, Y, \tilde{W}_1, \tilde{W}_2, \tilde{W}_3, \tilde{N}_{1,1}, \tilde{N}_{2,1}, \tilde{N}_{1,2}, \tilde{N}_{2,2}$, such that (3.2) and the following LMIs

$$\begin{bmatrix} \tilde{\Omega}_{1,1}^1 & \tilde{\Omega}_{1,2}^1 & \tilde{\Omega}_{1,3}^1 & \tilde{\Omega}_{1,8} \\ * & \tilde{\Omega}_{2,2}^1 & \tilde{\Omega}_{2,3}^1 & \tilde{\Omega}_{2,9} \\ * & * & \tilde{\Omega}_{3,3}^1 & 0 \\ * & * & * & -\vartheta \mathcal{S}\{X\} \end{bmatrix} < 0, \quad (3.27)$$

$$\begin{bmatrix} \tilde{\Omega}_{1,1}^2 & \tilde{\Omega}_{1,2}^2 & \tilde{\Omega}_{1,3}^2 & \tilde{\Omega}_{1,5} & \tilde{\Omega}_{1,8} \\ * & \tilde{\Omega}_{2,2}^2 & \tilde{W}_2^T & \tilde{\Omega}_{2,5} & \tilde{\Omega}_{2,9} \\ * & * & \tilde{\Omega}_{3,3}^2 & h\tilde{W}_3^T & 0 \\ * & * & * & -he^{-2\theta h}Q & -h\vartheta Y^T \\ * & * & * & * & -\vartheta \mathcal{S}\{X\} \end{bmatrix} < 0, \quad (3.28)$$

$$\begin{bmatrix} \tilde{\Omega}_{1,1}^3 & \tilde{\Omega}_{1,2}^3 & \tilde{\Omega}_{1,3}^3 & \tilde{\Omega}_{1,8}^3 \\ * & -\mathcal{S}\{\tilde{N}_{2,2}\} & G^T Y & \tilde{\Omega}_{2,9}^3 \\ * & * & (\varepsilon_2 - 1)M & 0 \\ * & * & * & -\vartheta \mathcal{S}\{X\} \end{bmatrix} < 0 \quad (3.29)$$

hold true, where matrix blocks such as $\tilde{\Omega}_{1,1}^1, \tilde{\Omega}_{1,2}^1$, and $\tilde{\Omega}_{1,3}^1$ are the same as those in Theorem 3.2 except that $\alpha = 1$ and $\bar{\alpha} = 0$.

4. Simulation example

In this section, we use a simplified inverted pendulum model [42] to illustrate the validity of the proposed NSETS-based controller design scheme under random deception attacks. This model can be

Table 1. Maximum exponential decay rate θ_{\max} under different probability uncertainty coefficient α_p .

α_p	0.06	0.07	0.08	0.09	0.1
θ_{\max}	0.5843	0.5459	0.5080	0.4708	0.4341

Table 2. Maximum exponential decay rate θ_{\max} under different estimated probability α .

α	0.88	0.86	0.84	0.82	0.8
θ_{\max}	0.4655	0.4613	0.4549	0.4459	0.4341

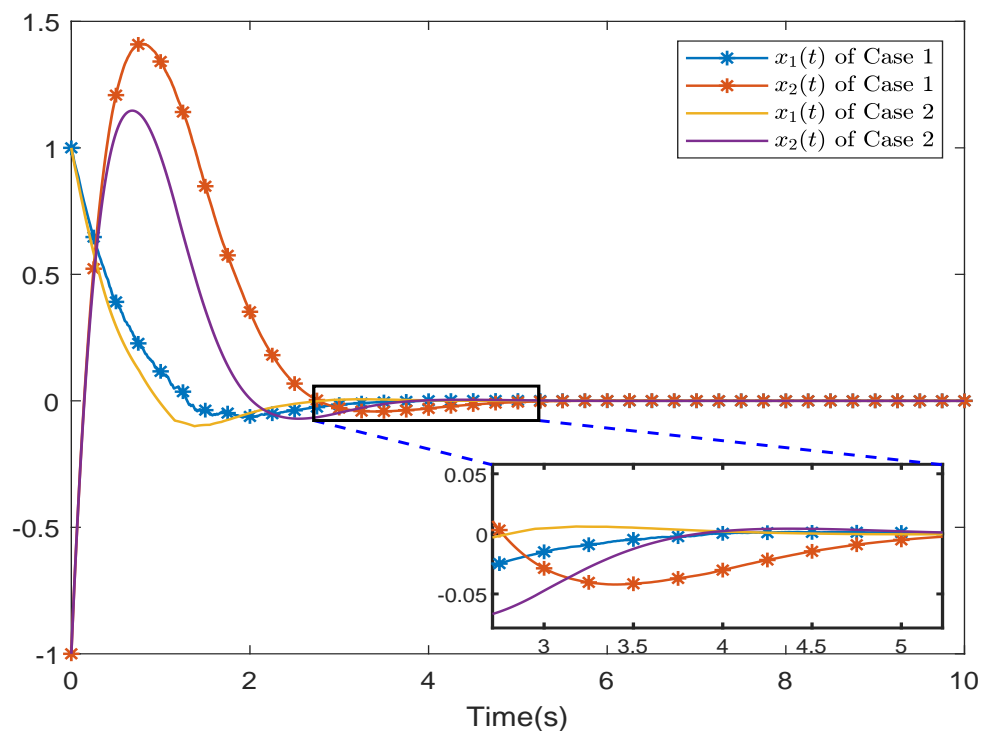


Figure 1. State trajectories.

described as follows:

$$\dot{x}(t) = \begin{bmatrix} -1.84 & 0.33 \\ 7.18 & -1.14 \end{bmatrix} x(t) + \begin{bmatrix} 2.43 \\ -0.42 \end{bmatrix} u(t).$$

First, we will co-design the triggering matrix W and control gain K to guarantee the MSE stability of the above system.

Choosing triggering thresholds $\varepsilon_1 = \varepsilon_2 = 0.1$, waiting interval $h = 0.05$, deception attack signal $d(x(t)) = \tanh(0.15x(t))$, uncertain probability term $\Delta\alpha(t) = \alpha_p \sin(t)$, and $\vartheta = 0.01$. Then, based on Theorem 3.2, when $\alpha = 0.8$, for different probability uncertainty coefficient α_p , the maximum exponential decay rate θ_{\max} can be obtained, which is listed in Table 1. It is straightforward to see that, as the probability uncertainty coefficient grows, the maximum exponential decay rate continues to

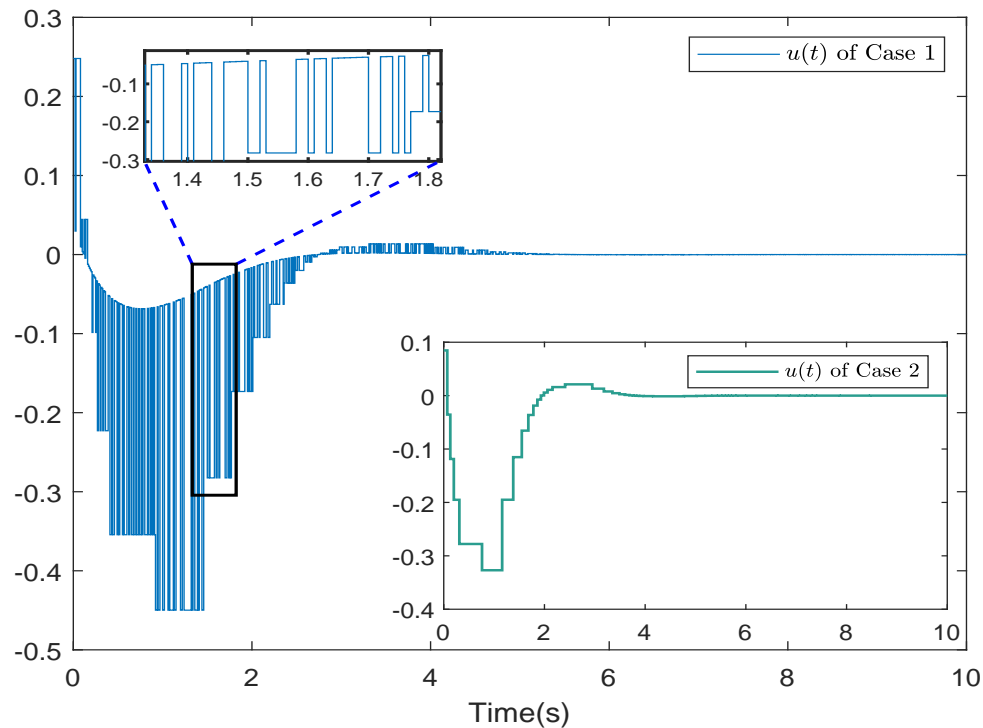


Figure 2. Control inputs.

deteriorate. In addition, Table 2 gives the values of θ_{\max} under the different probability α of deception attacks occurrence (when $\alpha_p = 0.1$). It can be found that as the value of α decreases (which means the frequency of deception attacks increases), the performance of the controller declines.

Next, to further reflect on the impact of deception attacks, we consider two cases: with and without deception attacks. In Case 1, the estimated probability of deception attacks and uncertainty coefficient are fixed as $\alpha = 0.5$, $\alpha_p = 0.08$, and other parameters are unchanged. By solving the LMIs in Theorem 3.2, the matrices M and K can be computed as

$$M = \begin{bmatrix} 0.7893 & -0.0158 \\ -0.0158 & 0.4791 \end{bmatrix}, \quad K = \begin{bmatrix} -0.0695 & -0.3175 \end{bmatrix}.$$

In Case 2, $\alpha = 1$ and $\alpha_p = 0$ (that is, the probability of occurring deception attacks is 0). By solving the LMIs in Corollary 2, the matrices M and K are calculated as

$$M = \begin{bmatrix} 1.2333 & 0.4199 \\ 0.4199 & 1.3964 \end{bmatrix}, \quad K = \begin{bmatrix} -0.1841 & -0.2688 \end{bmatrix}.$$

In the simulations, the initial condition is taken as $x_0 = [1, -1]^T$, and the running time is set as 10s. Figures 1–3 depict the state trajectories, control inputs, and release moment intervals between any two successively release moments of Cases 1 and 2, respectively. It can be seen that the state variables can converge to zero faster in the absence of deception attacks, which is consistent with the information in Table 2.

Finally, we will show the effect of triggering thresholds ε_1 and ε_2 in NSETS on the number of transmitted signals. Let t_n and t_r denote the number of trigger times and the ratio of transmitted signals,

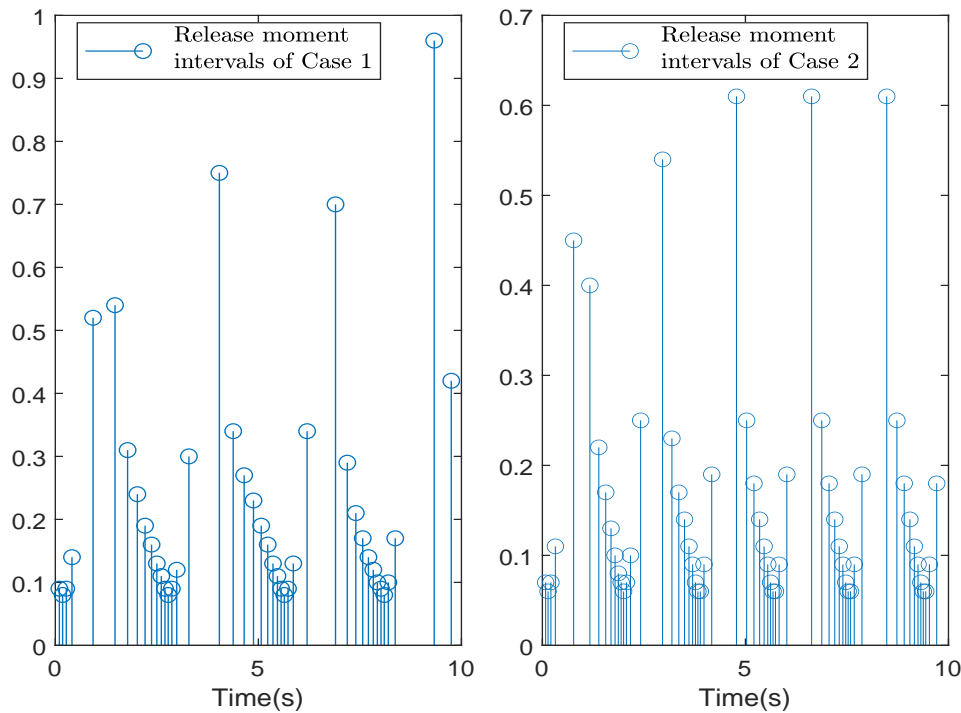


Figure 3. Release moment intervals.

Table 3. Comparison of trigger times under different triggering thresholds.

Scheme	TTS in [16]	SETS in [31]	NSETS in this paper			
$(\varepsilon_1, \varepsilon_2)$	(0, 0)	(0.1, 0)	(0.1, 0.05)	(0.1, 0.1)	(0.1, 0.15)	(0.1, 0.2)
t_n	200	71	54	43	36	32
t_r	100%	35.5%	27%	21.5%	18%	16%

respectively, and other parameters are the same as those in Case 1. Then, as shown in Table 3, when ε_1 and ε_2 are both set to 0, the NSETS degenerates into the TTS in [16], and t_n is as high as 200; when $\varepsilon_1 = 0.1$ and $\varepsilon_2 = 0$, the ETS changes into the SETS in [31], and t_n is reduced to 71; when ε_1 is fixed, t_n will continue to decrease as ε_2 increases. It can be observed that compared with SETS, the t_r of the NSETS decreases by more than 8.5%, and compared with TTS, the rate can be reduced by more than 73%. These results demonstrate that the addition of trigger condition $x^T(t_\sigma)Mx(t_\sigma)$ significantly conserves network resources while ensuring stability.

5. Conclusions

In this paper, the event-triggered exponential stabilization of NCSs under deception attacks has been investigated. Unlike existing ETSs, an NSETS, which additionally introduces a prescribed trigger term regarding the last triggering moment (i.e., $\varepsilon_2 x^T(t_\sigma)Mx(t_\sigma)$), has been designed in (2.4). It has been demonstrated that the additional trigger term plays a positive role in reducing the number of trigger times. Moreover, by using a Bernoulli variable with uncertain probability to characterize the randomness of deception attacks, a new mathematical model of random deception attacks has been

constructed in (2.3). A piecewise-defined Lyapunov function, which can make use of the system information at the instants t_σ and $t_\sigma + h$, has been established, which allows us to establish a sufficient exponential stability condition. Based on this, a co-design of the trigger and feedback-gain matrices under NSETS has been derived in terms of LMIs. Finally, a simulation example has been given to confirm the effectiveness of the developed design method. Future attention will be focused on the event-triggered security control problem of time-delay Markov jump systems [43–46].

Acknowledgments

This work was supported by the Key Research and Development Project of Anhui Province (Grant No. 202004a07020028).

Conflict of interest

The authors declare there is no conflict of interest.

References

1. Y. Song, Z. Wang, D. Ding, G. Wei, Robust model predictive control under redundant channel transmission with applications in networked DC motor systems, *Int. J. Robust Nonlinear Control*, **26** (2016), 3937–3957. <https://doi.org/10.1002/rnc.3542>
2. J. Liang, C. Gong, Y. Hou, M. Yu, W. Wang, Application of networked discrete event system theory on intelligent transportation systems, *Control Theory Technol.*, **19** (2021), 236–248. <https://doi.org/10.1007/s11768-020-00002-2>
3. L. Zhang, H. Gao, O. Kaynak, Network-induced constraints in networked control systems—A survey, *IEEE Trans. Ind. Inf.*, **9** (2013), 403–416. <https://doi.org/10.1109/TII.2012.2219540>
4. X. Liang, J. Xu, Control for networked control systems with remote and local controllers over unreliable communication channel, *Automatica*, **98** (2018), 86–94. <https://doi.org/10.1016/j.automatica.2018.09.015>
5. A. Kazemy, R. Saravanakumar, J. Lam, Master-slave synchronization of neural networks subject to mixed-type communication attacks, *Inf. Sci.*, **560** (2021), 20–34. <https://doi.org/10.1016/j.ins.2021.01.063>
6. J. Tian, R. Tan, X. Guan, Z. Xu, T. Liu, Moving target defense approach to detecting stuxnet-like attacks, *IEEE Trans. Smart Grid*, **11** (2019), 291–300. <https://doi.org/10.1109/TSG.2019.2921245>
7. H. Goyel, K. S. Swarup, Data integrity attack detection using ensemble based learning for cyber physical power systems, *IEEE Trans. Smart Grid*, (2022), In press. <https://doi.org/10.1109/TSG.2022.3199305>
8. Q. Wang, H. Yang, A survey on the recent development of securing the networked control systems, *Syst. Sci. Control Eng.*, **17** (2019), 54–64. <https://doi.org/10.1080/21642583.2019.1566800>

9. X. M. Zhang, Q. L. Han, X. Ge, D. Ding, L. Ding, D. Yue, et al., Networked control systems: A survey of trends and techniques, *IEEE/CAA J. Autom. Sin.*, **7** (2019), 1–17. <https://doi.org/10.1109/JAS.2019.1911651>
10. X. Jin, S. Lü, C. Deng, M. Chadli, Distributed adaptive security consensus control for a class of multi-agent systems under network decay and intermittent attacks, *Inf. Sci.*, **547** (2021), 88–102. <https://doi.org/10.1016/j.ins.2020.08.013>
11. Y. Yuan, H. Yuan, L. Guo, H. Yang, S. Sun, Resilient control of networked control system under DoS attacks: A unified game approach, *IEEE Trans. Ind. Inf.*, **12** (2016), 1786–1794. <https://doi.org/10.1109/TII.2016.2542208>
12. K. Paridari, N. OMahony, A. E. D. Mady, R. Chabukswar, M. Boubekour, H. Sandberg, A framework for attack-resilient industrial control systems: Attack detection and controller reconfiguration, *Proc. IEEE*, **106** (2017), 113–128. <https://doi.org/10.1109/JPROC.2017.2725482>
13. D. Du, C. Zhang, H. Wang, X. Li, H. Hu, T. Yang, Stability analysis of token-based wireless networked control systems under deception attacks, *Inf. Sci.*, **459** (2018), 168–182. <https://doi.org/10.1016/j.ins.2018.04.085>
14. Z. Hu, F. Deng, Y. Su, J. Zhang, S. Hu, Security control of networked systems with deception attacks and packet dropouts: A discrete-time approach, *J. Franklin Inst.*, **358** (2021), 8193–8206. <https://doi.org/10.1016/j.jfranklin.2021.08.015>
15. X. Gao, F. Deng, C. Y. Su, P. Zeng, Protocol-based fuzzy control of networked systems under joint deception attacks, *IEEE Trans. Fuzzy Syst.*, (2022), In press. <https://doi.org/10.1109/TFUZZ.2022.3194365>
16. T. Chen, B. A. Francis, *Optimal Sampled-Data Control Systems*, Springer Science & Business Media, (2012), 209–220. <https://doi.org/10.1007/978-1-4471-3037-6>
17. N. Gunasekaran, M. S. Ali, S. Arik, H. A. Ghaffar, A. A. Z. Diab, Finite-time and sampled-data synchronization of complex dynamical networks subject to average dwell-time switching signal, *Neural Networks*, **149** (2022), 137–145. <https://doi.org/10.1016/j.neunet.2022.02.013>
18. R. Vadivel, P. Hammachukiattikul, N. Gunasekaran, R. Saravanakumar, H. Dutta, Strict dissipativity synchronization for delayed static neural networks: An event-triggered scheme, *Chaos Solitons Fractals*, **150** (2021), 111212. <https://doi.org/10.1016/j.chaos.2021.111212>
19. N. Gunasekaran, Y. H. Joo, Robust sampled-data fuzzy control for nonlinear systems and its applications: Free-weight matrix method, *IEEE Trans. Fuzzy Syst.*, **27** (2019), 2130–2139. <https://doi.org/10.1109/TFUZZ.2019.2893566>
20. Z. M. Li, X. H. Chang, J. H. Park, Quantized static output feedback fuzzy tracking control for discrete-time nonlinear networked systems with asynchronous event-triggered constraints, *IEEE Trans. Syst. Man Cybern.: Syst.*, **51** (2021), 3820–3831. <https://doi.org/10.1109/TSMC.2019.2931530>
21. M. Dlala, S. O. Alrashidi, Rapid exponential stabilization of Lotka-McKendrick’s equation via event-triggered impulsive control, *Math. Biosci. Eng.*, **18** (2021), 9121–9131. <https://doi.org/10.3934/mbe.2021449>

22. P. Selvaraj, O. Kwon, S. H. Lee, R. Sakthivel, Equivalent-input-disturbance estimator-based event-triggered control design for master-slave neural networks, *Neural Networks*, **143** (2021), 413–424. <https://doi.org/10.1016/j.neunet.2021.06.023>
23. C. Ge, X. Liu, Y. Liu, C. Hua, Event-triggered exponential synchronization of the switched neural networks with frequent asynchronism, *IEEE Trans. Neural Networks Learn. Syst.*, (2022), In press. <https://doi.org/10.1109/TFUZZ.2022.3194365>.
24. C. Wang, Z. Ma, S. Tong, Adaptive fuzzy output-feedback event-triggered control for fractional-order nonlinear system, *Math. Biosci. Eng.*, **19** (2022), 12334–12352. <https://doi.org/10.3934/mbe.2022575>
25. D. Xu, Z. Li, G. Cui, W. Hao, Distributed fixed-time secondary control of an islanded microgrid via distributed event-triggered mechanism, *Int. J. Control*, (2022), In press. <https://doi.org/10.1080/00207179.2022.2032832>.
26. Z. Wu, J. Xiong, M. Xie, Improved event-triggered control for networked control systems subject to deception attacks, *J. Franklin Inst.*, **358** (2021), 2229–2252. <https://doi.org/10.1016/j.jfranklin.2020.12.018>
27. Y. Sun, J. Yu, X. Yu, H. Gao, Decentralized adaptive event-triggered control for a class of uncertain systems with deception attacks and its application to electronic circuits, *IEEE Trans. Circuits Syst. I Regul. Pap.*, **67** (2020), 5405–5416. <https://doi.org/10.1109/TCSI.2020.3027678>
28. J. Lian, Y. Han, Switching-like event-triggered control for networked Markovian jump systems under deception attack, *IEEE Trans. Circuits Syst. II Express Briefs*, **68** (2021), 3271–3275. <https://doi.org/10.1109/TCSII.2021.3065679>
29. B. Shen, Z. Wang, D. Wang, Q. Li, State-saturated recursive filter design for stochastic time-varying nonlinear complex networks under deception attacks, *IEEE Trans. Neural Networks Learn. Syst.*, **31** (2019), 3788–3800. <https://doi.org/10.1109/TNNLS.2019.2946290>
30. J. Cheng, Y. Wang, J. H. Park, J. Cao, K. Shi, Static output feedback quantized control for fuzzy Markovian switching singularly perturbed systems with deception attacks, *IEEE Trans. Fuzzy Syst.*, **30** (2021), 1036–1047. <https://doi.org/10.1109/TFUZZ.2021.3052104>
31. A. Selivanov, E. Fridman, Event-triggered \mathcal{H}_∞ control: A switching approach, *IEEE Trans. Autom. Control*, **61** (2016), 3221–3226. <https://doi.org/10.1109/TAC.2015.2508286>
32. J. Lunze, D. Lehmann, A state-feedback approach to event-based control, *Automatica*, **46** (2010), 211–215. <https://doi.org/10.1016/j.automatica.2009.10.035>
33. Z. Yan, X. Huang, J. Cao, Variable-sampling-period dependent global stabilization of delayed memristive neural networks based on refined switching event-triggered control, *Sci. China Inf. Sci.*, **63** (2020), 212201. <https://doi.org/10.1007/s11432-019-2664-7>
34. S. Ding, X. Xie, Y. Liu, Event-triggered static/dynamic feedback control for discrete-time linear systems, *Inf. Sci.*, **524** (2020), 33–45. <https://doi.org/10.1016/j.ins.2020.03.044>
35. Z. Yan, X. Huang, Y. Fan, J. Xia, H. Shen, Threshold-function-dependent quasi-synchronization of delayed memristive neural networks via hybrid event-triggered control, *IEEE Trans. Syst. Man Cybern.: Syst.*, **51** (2021), 6712–6722. <https://doi.org/10.1109/TSMC.2020.2964605>

36. W. Wu, L. He, J. Zhou, Z. Xuan, S. Arik, Disturbance-term-based switching event-triggered synchronization control of chaotic Lurie systems subject to a joint performance guarantee, *Commun. Nonlinear Sci. Numer. Simul.*, **115** (2022), 106774. <https://doi.org/10.1016/j.cnsns.2022.106774>
37. L. Xu, D. Xu, Mean square exponential stability of impulsive control stochastic systems with time-varying delay, *Phys. Lett. A*, **373** (2009), 328–333. <https://doi.org/10.1016/j.physleta.2008.11.029>
38. K. Gu, J. Chen, V. L. Kharitonov, *Stability of Time-Delay Systems*, Boston, MA: Birkhauser, 2003. <https://doi.org/10.1007/978-1-4612-0039-0>
39. K. Zhou, P. P. Khargonekar, Robust stabilization of linear systems with norm-bounded time-varying uncertainty, *Syst. Control Lett.*, **10** (1988), 17–20. [https://doi.org/10.1016/0167-6911\(88\)90034-5](https://doi.org/10.1016/0167-6911(88)90034-5)
40. J. Zhou, J. H. Park, Q. Ma, Non-fragile observer-based \mathcal{H}_∞ control for stochastic time-delay systems, *Appl. Math. Comput.*, **291** (2016), 69–83. <https://doi.org/10.1016/j.amc.2016.06.024>
41. E. Fridman, A refined input delay approach to sampled-data control, *Automatica*, **46** (2010), 421–427. <https://doi.org/10.1016/j.automat.2009.11.017>
42. Y. Wang, G. Yang, \mathcal{H}_∞ control of networked control systems with delay and packet disordering via predictive method, in *Proceedings of the 2007 American Control Conference*, (2010), 1021–1026. <https://doi.org/10.1109/ACC.2007.4282390>
43. J. Wang, Y. Zhang, L. Su, J. H. Park, H. Shen, Model-based fuzzy filtering for discrete-time Semi-Markov jump nonlinear systems using semi-markov kernel, *IEEE Trans. Fuzzy Syst.*, **30** (2022), 2289–2299. <https://doi.org/10.1109/TFUZZ.2021.3078832>
44. D. Tong, C. Xu, Q. Chen, W. Zhou, Y. Xu, Sliding mode control for nonlinear stochastic systems with Markovian jumping parameters and mode-dependent time-varying delays, *Nonlinear Dyn.*, **100** (2020), 1343–1358. <https://doi.org/10.1007/s11071-020-05597-4>
45. H. Shen, X. Hu, J. Wang, J. Cao, W. Qian, Non-fragile H_∞ synchronization for markov jump singularly perturbed coupled neural networks subject to double-layer switching regulation, *IEEE Trans. Neural Networks Learn. Syst.*, (2022), In press. <https://doi.org/10.1109/TNNLS.2021.3107607>.
46. M. S. Ali, N. Gunasekaran, R. Agalya, Y. H. Joo, Non-fragile synchronisation of mixed delayed neural networks with randomly occurring controller gain fluctuations, *Int. J. Syst. Sci.*, **49** (2018), 3354–3364. <https://doi.org/10.1080/00207721.2018.1540730>



AIMS Press

©2023 the Author(s), licensee AIMS Press. This is an open access article distributed under the terms of the Creative Commons Attribution License (<http://creativecommons.org/licenses/by/4.0>)